Show that the following circuit works as a demodulator of frequency modulation (FM) signal (quadrature demodulator).



Here the phase shifter gives the shift proportional to the frequency difference between input and the career frequency  $\omega_0$ . The shift at  $\omega_0$  is  $\pi/2$  as shown in the right (this can be achieved with resonant circuits). The low-pass filter cuts components with frequencies as high as  $\omega_0$ .



(hint) Assume the original signal f(t) is much slower than the carrier  $A \cos(\omega_0 t)$ . Then the input can be approximated as

$$s(t) = A\cos\{[\omega_0 + k_f f(t)]t\}.$$

Then the phase shifter output is  $q_{\rm ps}(t) = A \sin\{[\omega_0 + k_f f(t)]t - ak_f f(t)\}.$ 

Taking product and high-frequency filtering gives ... (use  $ak_f f(t) \ll 1$ ).

In the following phase lock loop (PLL) circuit, the initial (t = 0) oscillation frequency of voltage-controlled oscillator (VCO)  $\omega$  deviates from  $N\omega_0$  by  $\Delta\omega$ . Obtain the relaxation time of  $\omega$  to  $N\omega_0$ .



(hint) Here we can put  $\theta_i = 0$  hence input  $= V_i \sin \omega_0 t$  without loosing generality. Similarly output  $= V_o \sin[N\omega_0 t + \theta_o(t)]$ . Now  $\omega = N\omega_0 + d\theta_o/dt$  and it is easy to write  $d\theta_o/dt$  with A,  $G_p, V_c, \theta_o(t)$  and a constant.

Solve the difference equation below with z-transform.

$$\begin{cases} x(n) - 2x(n-1) = n & (n \ge 0) \\ x(n) = 0 & (n < 0) \end{cases}$$

(hint) z-transform of *n* is  $\frac{z}{(z-1)^2}$  as in the table (slide no.14).

Then z-transform of x(n) : X(z) is easily obtained. Inverse z-transform gives x(n).

Answer sheet submission deadline: 11<sup>th</sup> Jan. 2017.